LB8632V



Low-Voltage/Low-Saturation Camera Motor Driver

Overview

The LB8632V is a general-purpose camera motor driver IC that supports low-saturation output and low-voltage drive and thus can be used in a wide range of applications. The LB8632V is a miniature thin form-factor IC that provides circuit structures and I/O logic that reduce development times and costs and support reduced lot sizes and end product diversity, despite increasing miniaturization and functionality in application products.

Since the LB8632V supports low-voltage operation (starting from 1.2 V) in addition to low-saturation outputs, it provides characteristics that can withstand operation in low-voltage ranges even in 3 V popularly-priced cameras that do not include a step-up circuit. Since the LB8632V supports IIL, control logic can be optimized by using two ICs in parallel, thus allowing even more actuators to be driven with a small number of CPU output ports. Thus the LB8632V can also be used in 6 V top-of-the-line end products.

Features

• 3, 6 V and a wide application range: from popularly priced models to top-of-the-line products

The external transistors and the number of ICs (one or two) is determined by the number of IC required motors and actuators.

Support for motor standby, forward, reverse and braking control for four motors using only five CPU port lines. This requires two external transistors and two ICs. The LB8632V also provides non-operating mode logic for the operating control signals (the IN input) so that ICs with other functions can be used on the same ports.

• Built-in 1 CH low-voltage drive low-saturation, forward/reverse motor drive

 I_O peak = 3.0 A maximum (t \leq 100 ms: single shot, between OUT1 and OUT2)

 $I_{O} DC = 1.0 A$ maximum

 V_{O} sat = 0.45 V typical (at I_{O} = 1 A)

Operating voltage range: $V_B = 1.2$ to 7.5 V ($V_B = V_S = V_{DD}$ = battery voltage)

• Built-in low-voltage drive - 0.5 CH forward/reverse motor driver

Supports 1.5 CH bridge operation with the use of two external transistors. Also supports LED or solenoid drive with constant-current drive.

Operating voltage range: $V_B = 1.2$ to 7.5 V ($V_B = V_S = V_{DD}$ = battery voltage)

• Built-in regulator predriver

 V_{CC} can be regulated at 1.86 V by connecting the V_{CC} control pin (V_{CONT}) to the reference voltage, V_{REF} (1.36 V). Also, the V_{CC} output voltage can be varied by inputting the CPU D/A output to V_{CONT} .

Either the IC itself or only the $V_{\rm CC}$ regulator can be turned on according to the operating mode of each motor.

In addition, this function can also be used as a battery check comparator.

Operating voltage range: $V_B = 1.5$ to 7.5 V

Three independent power supply line systems

The LB8632V supports either single supply specifications ($V_B = V_S = V_{DD}$) or dual supply specifications (battery/step-up supply) by providing a power supply voltage pin (V_B), a motor supply pin (V_S), and a CPU interface pin (V_{DD}). Also, motor output can be stabilized by taking the motor power supply from the V_{CC} regulator.

Support for both CMOS and n-channel open drain outputs from the CPU.

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- Built-in thermal protection circuit This circuit limits the output current if the IC overheats due to excessive loading or an output short and thus prevents the destruction of the IC.
- Provided in the miniature ultrathin SSOP-20 package (6.4 × 6.5 × 1.6 mm)



Absolute Maximum Ratings at $Ta = 25^{\circ}C$

Package Dimensions

unit: mm

3179A-SSOP20



Parameter	Symbol	Conditions	Ratings	Unit
	V _B max		8.0	V
Maximum supply voltage	V _S max		8.0	V
	V _{DD} max		8.0	V
Output current	I _O max	Between OUT1 and OUT2 (t \leq 100 ms, single pulse)	3.0	A
	V _{OUT} 1	OUT1, OUT2	V _S + V _F	V
Output supply voltage	V _{OUT} 2	OUT3N, OUT3P	V _S	V
	V _{OUT} 3	V _{REF}	V _B	V
Input supply voltage	V _{IN} 1	MD0 to MD2, IN1, IN2	V _{DD}	V
	V _C	V _{CONT}	VB	V
Allowable power dissipation	Pd max	Mounted on a 50 \times 50 \times 1.5 mm^3 glass-epoxy printed circuit board	800	mW
Operating temperature	Topr		-20 to +75	°C
Storage temperature	Tstg		-55 to +150	°C

Parameter	Symbol	Conditions	min	typ	max	Unit	Note
Standby current	I _{B STB}	V _B = 7.5 V		0.1	1	μΑ	1
[DC motor system]							
Operating voltage range 1	V _B 1	V _B system	1.2		7.5	V	2
Operating voltage range r	V _S 1	V _B system	1.0		VB	V	3
Output saturation voltage	Vsat	$I_{O} = 600 \text{ mA}, V_{B} = V_{S} = V_{DD} = 1.8 \text{ V}$		0.30	0.45	V	4
(pnp + npn)	Vsat	I _O = 1000 mA		0.45	0.65	V	4
Lower side output saturation voltage	VsatN	I _O = 600 mA		0.15	0.25	V	24
Spark killer diode forward voltage	V _F	I _F = 600 mA		0.9	1.4	V	25
	I _B +	OUT3P	-15	-20	-28	mA	5
Output constant current	I _B -	OUT3N	15	20	28	mA	5
Operating current drain 1	I _B 1	V _B + V _S (maximum forward/reverse/ braking drive)		80	98	mA	6
Operating current drain 2	I _B 2	V _B + V _S (standby in drive mode)		5.0	8.0	mA	7
Operating current drain 3	I _B 3	$V_B + V_S$ (maximum single side drive)		50	70	mA	8

Parameter	Symbol	Conditions	min	typ	max	Unit	Note	
[V _{REF} system]								
Operating voltage range 2	V _B 2		1.5		7.5	V	9	
V _{CONT} input voltage range	V _{CR}		0		V _B	V	10	
V _{REF} voltage	V _{REF}	I _{REF} = 10 μA	1.29	1.36	1.43	V	11	
I/O voltage ratio	O _{IR}	V _{CC} /V _{CONT}	1.35	1.37	1.39	Times	12	
Minimum constant output voltage	V _{CC} min			0.8	0.9	V	13	
Regulator voltage	V _{REG}	$V_{\text{REF}} = V_{\text{CONT}}, C2 = 10 \ \mu\text{A}$	1.76	1.86	1.96	V	14	
I _B pin sink current	I _{IB}		8.0			mA	26	
Line regulation	ΔV_{OLN}	$2.0 \text{ V} \le \text{V}_{\text{B}} \le 7.5 \text{ V}$		10	20	mV	15	
Load regulation	ΔV_{OLD}	$50 \text{ mA} \le I_{CC} \le 500 \text{ mA}$		20	50	mV	16	
V _{CONT} input current	I _{VC}				1	μA	17	
Operating current drain 4	I _B 4	V _B system		5.0	8.0	mA	18	
[V _{DD} control input system]		·			•			
V _{DD} voltage range	V _{DR}		1.2		7.5	V	19	
Control pip input current	I _{INL}	$V_{INL} = GND, V_{DD} = 3 V$		-120	-150	μA	20	
	I _{INH}	$V_{INH} = V_{DD}, V_{DD} = 3 V$	-3		0	μA	20	
Input low-level voltage	V _{IL}		-0.3		V _{DD} - 1.0	V	21	
Input high-level voltage	v_{IH}		V _{DD} - 0.3		V _{DD}	V	21	
Operating current drain	I _{VD}	MD0, MD1, MD2, IN1 and IN2 = GND, $V_{DD} = 5 V$	1.7	2.2	2.7	mA	22	
Thermal protection operating temperature	T _{SD}	Design target value		180		°C	23	

Electrical Characteristics at Ta = 25° C, V_B = V_S = V_{DD} = 3.0 V

Note: There are no limitations on the magnitude relationships between the V_B, V_S and V_{DD} supply voltages.
1. Stipulates the total leakage current for V_B = V_S = V_{DD} when the IC is in standby mode.
2. Stipulates the operating range voltages (for guaranteed functionality) when a single power supply with V_B = V_S = V_{DD} is used for the DC motor system.
3. Stipulates the operating range voltages (for guaranteed functionality) for the DC motor system V_S pin, i.e., the motor power supply. The drive

current will be a constant current when $V_B = V_S = 2.0$ V or higher.

4. Stipulates the output saturation voltage when either the DC motor driver $V_B = V_S = V_{DD} = 1.8$ V and the output current is 600 mA or when $V_B = V_S$ = V_{DD} = 3.0 V and the output current is 1000 mA.

5. Stipulates the current emitted or accepted by the OUT3P and OUT3N pins. These are constant currents as long as $V_B = V_S = 2.0$ V or higher.

6. Stipulates the maximum total current drain for the V_B and V_S pins for the forward, reverse, or brake operations when drive between the DC motor system OUT1 and OUT2 outputs is used.

7. Stipulates the maximum total current drain for the V_B and V_S pins for the DC motor driver drive mode standby state.

Stipulates the maximum total current drain for the V_B and V_S pins for the DC motor driver single-sided drive mode, i.e. when a single output from 8. the OUT1, OUT2, OUT3P and OUT3N pins is used.

Stipulates the operating voltage range (for guaranteed functionality) for the V_{REF} circuit system. 9.

10. Stipulates the input voltage range for the Vcont pin. When a constant-voltage circuit is formed by adding an external transistor, the voltage is held constant by sensing the V_{CC} pin.

11.Stipulates the reference voltage generated at the V_{RFF} pin.

12. Stipulates the ratio of the constant V_{CC} pin voltage value to the V_{CONT} pin input voltage. 13. Stipulates the minimum output voltage for the V_{CC} pin constant voltage output. The voltage will only fall to this value, even when the V_{CONT} pin is set to 0 V

14. Stipulates the V_{CC} pin stabilized output voltage when the V_{REF} and V_{CONT} pins are connected directly. 15. Stipulates the change in the value of the V_{CC} voltage when V_{CC} is set for constant voltage output and the V_B voltage varies from 2 V to 7.5 V. 16. Stipulates the change in the value of the V_{CC} voltage when V_{CC} is set for constant voltage output and the load current varies from 50 mA to 500 mΑ

17. Stipulates the V_{CONT} pin input current in the range 0 V ≤ V_{CONT} ≤ V_B − 0.3 V. The value of the V_{CC} output constant voltage can be varied by inputting the CPU D/A output.

18.Stipulates the V_B system maximum current drain in all modes when the V_{CC} output has no load.

19. Stipulates the operating voltage range (for guaranteed functionality) for the V_{DD} pin. The V_{DD} pin is connected either to the CPU power supply or V_B . 20. Stipulates the input current and allowable leakage current for the control input pins: MD0, MD1, MD2, IN1 and IN2.

21. Stipulates the high and low input voltages for the control input pins: MD0, MD1, MD2, IN1 and IN2. (When input pins are open, they appear to be high-level inputs.)

22. Stipulates the V_{DD} pin current drain when all the for the control input pins (MD0, MD1, MD2, IN1 and IN2) are at the ground level.

23. When the temperature exceeds the stipulated temperature, output current limitation is applied, thus protecting the IC. The stipulated temperature is a design target value and is not tested prior to shipment.

24. Stipulates the lower side output saturation voltage in the OUT1 and OUT2 braking modes (when In = 600 mA).

25. Stipulates the OUT1 and OUT2 spark killer diode forward voltage (when I_F = 600 mA).

26. Stipulates the minimum value of the I_B pin sink current.





Note: Connect both P-GND pins and both V_S pins. **A04089**

	MD IN		1D IN			OUT		V	Nete				
0	1	2	1	2	0011	0012	3P 3N	3P V _{CC} 3N		Note			
н	н	н	—	-	—	_	—	—	Standby (zero cu	urrent drain)			
		Н Н — — —	Standby										
			н	L	L	н]	07	Forward	Mator 1			
			L	н	н	L] —	On	Reverse				
			L	L	L	L]		Brake				
н	L	н	—	—	—	—	—	On	The IN input is ig	gnored. Motor 3 drive used.			
		L				Н	Н	—	—			Standby	
	LL		Н	L	—	Н]	On Forward Reverse Brake	Forward	Motor 2			
			L	Н	—	L			Reverse Brake				
			L	L	—	L							
	L	L	—	—	—	—	—	On	Only V_{CC} on				
				Н	Н	—	—	—		Standby			
	ц .		Н	L	L	—	3P		Forward	Motor 4			
			L	н	Н	—	3N		Reverse	External transistor			
			L	L	L	—	3N		Brake				
L	L	н	_	—	—	_	—	On	The IN input is ig	gnored			
			н	н	—	—	—		Standby				
			н	L	—	_	3P		Forward	Motor 5			
			L	н	_	_	3N		Reverse	External transistor			
			L	L	—	—	3N		Brake				

Truth Table (active low)

Note: The "—" entries for active-low/IN inputs are don't care states, and the "—" entries for OUT outputs are OFF states. See the sample application circuit for the motor number.

Internal Block Diagram



Wiring Notes

- 1. Connect both the P-GND and both the V_S pins. Although both the P-GND and V_S lines are connected internally, both must be connected to provide currents of 1 A or over, or to provide even lower saturation output. However, operation with only one of each of these pairs connected, or with through power supply wiring, is possible.
- 2. Since large currents flow in the V_S and P-GND lines, these lines should be made thicker, and line impedance reducing capacitors should be inserted in the vicinity of the IC.
- 3. Since S-GND is the ground for the control system, rather than using the same wiring as the P-GND line, it is preferable to connect this pin to the CPU ground line.
- 4. If the CPU outputs are CMOS outputs, connect V_{DD} to the CPU power supply line and if they are n-channel open drain outputs, connect V_{DD} to the V_B pin (battery). However, since V_{DD} is the control input system power supply, it should not be set to the same impedance as the V_S line.

Sample OUT3P/OUT3N Pin Applications

The OUT3P and OUT3N pins support -20 mA (typical) and 20 mA (typical) constant current drive, respectively. Constant current is supported when $V_B = V_S = 2.0 \text{ V}$ or higher. A 5 k Ω shunt resistor between the OUT3P pin and ground is included within the IC. Inversely, the OUT3N pin is an open collector pin, and there is no resistor inserted between this pin and the V_S pin. A current limiting resistor (resistance maintaining transistor) may be inserted between the OUT3P and OUT3N pins and the external transistors. These pins can also be used for direct LED drive, battery check, or other power supply switching functions.



LED Resistance maintaining transistor (PNP) Resistance maintaining transistor (NPN) Power supply switch for battery check or other function **A04091**

Recommended Transistors

Low-saturation transistors		
2SB815/2SD1048	СР	$I_0 = 0.7 A$
2SB1120/2SD1620	PCP	$I_0 = 2.5 \text{ A}$
Transistors with spark kill	er diodes and E	E-B resistors
CPH3120, 2SD2324	СР	$I_0 = 0.8 A$
2SB1397/2SD2100	PCP	$I_0 = 2 A$
Transistors with resistors		
2SA1520/2SC3914	СР	$I_0 = 0.5 A$

Sample I_B and V_{CC} Pin Applications

1. Stabilized Power Supply

A low-saturation type stabilized power supply can be formed using an external pnp transistor (2SB815). This circuit has a wide operating voltage range of 1.5 to 7.5 V, and can supply stable power to other pins and loads. This circuit outputs a constant voltage 1.37 times the Vcont input voltage. A 1.86 V constant voltage output can be acquired by directly connecting the V_{REF} reference voltage (1.36 V) to the Vcont pin.

 This circuit sets up a constant voltage output of 1.86 V or lower. When R1 is 20 kΩ, the output voltage can be set to any voltage between 0.8 and 1.86 V (typical) by adjusting R2.



 This circuit sets up a constant voltage output of 1.86 V or higher. Taking the manufacturing variation in the IC internal sensing resistor (3.6 to 10 kΩ) into account, the total resistance of the external sensing resistors (R1 and R2) should be about 1 kΩ. Example: A constant voltage output V_{REG} of 3.05 V is acquired when R1 is 360 Ω and R2 is 680 Ω.



• Making the OUT pin output voltage a constant voltage. Connect the OUT pin, which is to be made a constant voltage output, to the V_{CC} pin. When one side of the bridge output is connected, when the connected side is a high-level output the circuit will be a constant voltage drive circuit, and when the other side is a high-level output the circuit will be a saturated drive circuit.

This circuit uses a resistance maintaining transistor to lower the

A 0.8 V (typical) hold voltage is acquired when R2 is 0 Ω .

• Generating a hold voltage.

V_{CONT} input and generate a hold voltage.







• Generating a set constant voltage for different modes. The V_{CONT} input has a high impedance of 1 μ A maximum, and thus the set constant voltage can be changed as required for each mode by inputting the CPU D/A converter output.

2. Battery Check

Tr2 will turn off when the Vcont input voltage, which is $V_B \times \frac{r2}{r2 + r1}$, falls to 0.95 V or lower.

Example: This circuit can check for V_B being 2.0 V or lower by setting r1 to be 22 k Ω and r2 to be 20 k Ω . Note that the Tr3 transistor can also use the OUT3N output.



3. Light Measurement

Tr2 will turn on when the V_{CONT} input voltage becomes $V_B \times 70\%$ (= 9 kΩ/(3.6 kΩ + 10 kΩ)) or higher. Note that the Tr3 transistor can also use the OUT3N output.



A04100

Power Supply Specifications

1. Single Supply Specifications



Since the V_{DD} line is the control input system line, it should not have the same impedance as the V_S line.

2. Step-Up Supply Specifications



 $\begin{array}{l} Voltage \ application \ conditions \\ V_{CONT} \leq V_B \\ V_{IN}, V_{MD} \leq V_{DD} \\ There \ are \ no \ limitations \ on \ the \ magnitude \\ relationships \ between \ the \ V_B, \ V_S \ and \ V_{DD} \\ eucply \ voltage \end{array}$

supply voltages.







MD		Modo		
0	2	IVIOGE		
н	н	Standby (zero current drain)		
н	L	Motor 1		
L	Н	Solenoid, battery check		



MD	Mada
2	Mode
Н	Standby (zero current drain)
L	Motor 1 (single-direction regulator output)



MD		Mode	
0	2	Mode	
Н	Н	Standby (zero current drain)	
Н	L	Motor 1 (single-direction regulator output)	
L	Н	V_S line saturated output (battery voltage switch)	
L	L	Motor 4 (single-direction regulator output)	



MD		Modo	
0	2	Mode	
Н	н	Standby (zero current drain)	
Н	L	Motor 1	
L	н	Only the V _{CC} regulator output on	
L	L	Motor 4 (single-direction regulator output)	



The IC2 preamplifier can be used to implement the battery check circuit shown in Sample Application 1.

A04108

M	ID	Mode	
1	2	Nidde	
н	Н	Standby (zero current drain)	
н	L	Motor 1	
L	Н	Motor 3 (bidirectional regulator output)	
L	L	Motor 2 (single-direction regulator output)	



MD			Mada
0	1	2	Wode
Н	Н	н	Standby (zero current drain)
н	Н	L	Motor 1
н	L	н	Motor 3 (bidirectional regulator output)
н	L	L	Motor 2 (single-direction regulator output)
L	L	L	Only the V _{CC} regulator output on
L	L	н	Motor 4 (bidirectional regulator output)
L	Н	Н	IRED, LED



MD			Mada
0	1	2	Widde
Н	Н	н	Standby (zero current drain)
Н	Н	L	Motor 1 (bidirectional regulator output)
Н	L	н	Motor 3 (bidirectional regulator output)
Н	L	L	Motor 2 (bidirectional regulator output)
L	L	L	Only the V _{CC} regulator output on
L	Н	Н	Motor 5 (H bridge formed using external transistors)

Truth Table for Two LB8632V ICs

This truth table specifies the logic when using two LB8632V ICs (IC1 and IC2). As shown in the figure below, the five lines from the CPU should be connected to MD0, MD1, MD2, IN1, and IN2 for IC1 and to MD0, MD2, MD1, IN2, and IN1 for IC2, respectively.



Note: The "—" entries for active-low/IN inputs are don't care states, and the "—" entries for OUT outputs are OFF states. See the sample application circuit for the motor number.

MD			IN		IC1				IC2						
0	1	2	1	2	OUT1	OUT2	OUT3 PNP	OUT3 NPN	OUT2	OUT1	OUT3 PNP	OUT3 NPN	V _{CC}	Note	
н	н	н	-	—	—	_	—	—	—	—	_	—	-	Standby (zero current drain)	
н	н	L	н	н	—	—	—	—	—	—	—	—	On	Standby Forward Reverse Brake	Motor 1
			н	L	L	Н	—	—	—	—	—	—			
			L	Н	н	L	—	—	—	—	—	—			
			L	L	L	L	—	—	—	—	—	—			
	L	н	н	Н	—	—	—	—	—	—	—	—	- On	Standby Forward Reverse Brake	Motor 3
			н	L	—	—	—	—	L	н	—	—			
			L	Н	—	—	—	—	н	L	—	—			
			L	L	—	—	—	—	L	L	—	—			
	L	L	н	Н	—	—	—	—	—	—	—	—	- On	Standby Forward Reverse Brake	Motor 2
			н	L	—	Н	—	—	L	—	—	—			
			L	Н	—	L	—	—	н	—	—	—			
			L	L	—	L	—	—	L	—	—	—			
L	L	L	—	—	—	—	—	—	—	—	—	—	On	Only V _{CC} on	
	н	L	н	Н	—	—	—	—	—	—	—	—	- On		
			н	L	L	—	On	—	—	—	—	—			
			L	Н	н	—	—	On	—	—	—	—			
			L	L	L	—	—	On	—	—	—	—			
	L	н	н	Н	—	—	—	—	—	—	—	—	- On	Standby Forward I Reverse B Brake	Motor 4 External transistor
			н	L	—	—	—	—	—	н	—	On			
			L	Н	—	—	—	—	—	L	On	—			
			L	L	—	—	—	—	—	L	—	On			
	Н	н	Н	Н	-	—	_	—	_	_	—	-	- On	Standby Forward Reverse Brake	Motor 5 External transistor
			Н	L	-	—	On	—	-	_	-	On			
			L	Н	-	—	_	On	—	—	On	_			
			L	L	_	_	_	On	_	_	_	On			







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